

#### ECE 344

### **Microwave Fundamentals**

Assistant Professor Dr. Gehan Sami

# Example 1(2.2 in the book)

Z<sub>0</sub>=100, z<sub>L</sub>= 0.4+j0.7 (Zl=40+j70) l=.3λ Find  $\Gamma_L$ , Γ(-*l*), Zin, SWR, Return loss

From smith chart We can find: From reflection coefficient E scale  $|\Gamma_{\rm L}| = .59.\theta = 104^{\circ}$ From SWR se 21.0 61.0 86.0 12:0 SWR=3.8 From return l RADIALLY SCALED PARAMETERS SWR=3.8 TOWARD GENERATOR Scale, return ∞10040 20 10 1.2 1.1 1 15 1.6 1.8 2 1.2 1.3 1.4 loss=4.6dB NBM COEFF. Ear. 0.1 0.2 1.4 1.5 1.6 1.7 1.8 1.9 2 0.05 0.01 1.1 0.9 0.8 0.1 1.2 1.3 3  $,\Gamma(-l)=.59L^{-1}$ 0.5 0.4 0.3 0.2 0.1 0.99 0.6 0.5 0.4 0.3 0.2 0.1 0.9 0.95 0.9 0.7  $|\Gamma_{L}| = .59.$ CENTER 0.6 0.7 1 1.1 1.2 1.3 1.4 1.5 1.6 1.7 1.8 1.9 0.3 0.8 0.9 0.2 0.1 ORIGIN

0.105λ



 $z_{\rm L} = 0.4 + j0.7$ 

SWR circle



#### **Effect of moving TWG** $\lambda/4$

$$z_{in}(l = \lambda/4) = \frac{1}{z_L} = y_L$$
 where  $z_{in} = \frac{Z_{in}}{z_0}, \quad z_L = \frac{Z_L}{z_0}, \quad y_L = \frac{Y_L}{y_0}$ 

the length  $\lambda/4 \rightarrow \text{moving } \pi$  on smith chart  $\rightarrow$  gives us admittance of the load



A transmission line of length  $l = \lambda / 4$ 

transforms a load reflection  $\varGamma_{\rm load}$  to its input as

$$\Gamma_{in} = \Gamma_{load} \ e^{-j2\beta l} = \Gamma_{load} \ e^{-j\pi} = -\Gamma_{load}$$

Thus a normalized load impedance z is transformed into 1/z.

In particular, a short circuit at one end is transformed into an open circuit at the other. This is a particular property of the  $\lambda/4$  transformers.

when adding a transmission line to some terminating impedance we move clockwise through the Smith-Chart.



### Admittance (Y) Calculations

Note:

$$Y(-\ell) = \frac{1}{Z(-\ell)} = \frac{1}{Z_0} \left( \frac{1 - \Gamma(-\ell)}{1 + \Gamma(-\ell)} \right)$$

$$=Y_0\left(\frac{1+\left(-\Gamma\left(-\ell\right)\right)}{1-\left(-\Gamma\left(-\ell\right)\right)}\right) \qquad Y_0=\frac{1}{Z_0}$$

$$\Rightarrow Y_n(-\ell) = \frac{Y(-\ell)}{Y_0} = \left(\frac{1 + (-\Gamma(-\ell))}{1 - (-\Gamma(-\ell))}\right) = G_n(-\ell) + jB_n(-\ell)$$

Define:  $\Gamma' = -\Gamma$   $Y_n(-\ell) = \left(\frac{1+\Gamma'}{1-\Gamma'}\right)$ Contained Same mathematical form as for  $Z_n$ :  $Z_n(-\ell) = \left(\frac{1+\Gamma}{1-\Gamma}\right)$ 

Conclusion: The same Smith chart can be used as an admittance calculator.

# Admittance (Y) Chart

As an alternative, we can continue to use the original  $\Gamma$  plane, and add admittance curves to the chart.

$$Y_{n}\left(-\ell\right) = \left(\frac{1+\left(-\Gamma\left(-\ell\right)\right)}{1-\left(-\Gamma\left(-\ell\right)\right)}\right) = G_{n}\left(-\ell\right) + jB_{n}\left(-\ell\right)$$

Compare with previous Smith chart derivation, which started with this equation:

$$Z_{n}\left(-\ell\right) = \left(\frac{1+\left(\Gamma\left(-\ell\right)\right)}{1-\left(\Gamma\left(-\ell\right)\right)}\right) = R_{n}\left(-\ell\right) + jX_{n}\left(-\ell\right)$$

If  $(R_n X_n) = (a, b)$  is some point on the Smith chart corresponding to  $\Gamma = \Gamma_0$ , Then  $(G_n B_n) = (a, b)$  corresponds to a point located at  $\Gamma = -\Gamma_0$  (180° rotation).

$$\Rightarrow R_n = a \text{ circle, rotated 180°, becomes } G_n = a \text{ circle.}$$
  
and  $X_n = b \text{ circle, rotated 180°, becomes } B_n = b \text{ circle.}$ 

Side note: A 180° rotation on a Smith chart makes a normalized impedance become its reciprocal.

# Admittance (Y) Chart (cont.)



# Impedance and Admittance (*ZY*) Chart



Short-hand version

 $R_n = 1$ 

 $X_n = -1$ 

 $\Gamma$  plane



#### **Using impedance-Admittance Smith Chart**



#### **Adding Elements**

# Admittance

A matching network is going to be a combination of elements connected in series AND parallel.

Impedance is well suited when working with series configurations. For example:

$$V = ZI \qquad \qquad Z_L = Z_1 + Z_2$$

Impedance is NOT well suited when working with parallel configurations.

$$Z_{\mathrm{L}} = \frac{Z_1 Z_2}{Z_1 + Z_2}$$





For parallel loads it is better to work with admittance.

$$\mathbf{I} = \mathbf{Y}\mathbf{V} \qquad \mathbf{Y}_1 = \frac{\mathbf{I}}{Z_1} \qquad \mathbf{Y}_L = \mathbf{Y}_1 + \mathbf{Y}_2$$







Adding Parallel (Shunt) Components

→ <u>Admittance Smith Chart</u>



### **Series and Shunt Elements**



Note: The Smith chart is not actually being used as a transmission-line calculator but an impedance/admittance calculator. Hence, the normalizing impedance is arbitrary.

### **Adding a Series Capacitor**



Z=0.5+j0.7

If we have initial impedance Z=0.5+j0.7 We add a series capacitor

Since resistance does snot change We move on constant circle from j0.7 to -j0.3

Z=0.5-j0.3

Values on ADS for f=1GHz Z=25+j35 Zin=25-j15 1/wc=Zin-Z=50 so C=1/(50\*2pi)=3.18pF



### **Adding a Series Inductor**



If we have Z=0.5-j0.4 We add series inductor jX=j1.4

We move on resistance circle Zin=0.5+j1.0

Values on ADS Z=25-j20 Zin=25+j50 jwL=j70 so L=70/(2pi)=11.14nH



### **Adding a Shunt Capacitor**



For ZL=1+j1.0 On admittance chart YL=0.5-j0.5

Adding shunt capacitor With JB=j1

Yin=0.5+j0.5

Read from impedance chart

Zin=1-j1.0



### **Adding Shunt Inductor**



YL=0.2+j0.5

τ Γ Γ

adding shunt inductor with JB= - j0.7

Yin=0.2-j0.2

Read from impedance chart Zin=2.4+j2.5

#### Example 6

what is the input impedance of network shown in Fig below



Solution : use impedance admittance smith chart

ZL			
Z: 1.000	00 +j	1.0	0000
Y: 0.500	00 +j	-0.5	50000
Z1,	Y1		
Z:	0.55699	+j	0.89652
Y:	0.50000	+j	-0.80479
Z2,Y2			
z:	0.55699	+j	-0.48811
Υ:	1.01549	+j	0.88992
	Z3,Y3		
Z:	0.20511	+j	-0.39989
Y:	1.01549	+j	1.97981
Zin, Yin			
Z:	0.20485	+j	0.49905
Y:	0.70391	+j	-1.71486

